

Claims

- 5 **1.** A method for estimating the phase in a digital communication system comprising the steps of:
- receiving and storing a block of observations Y_k ;
 - executing at least one phase locked loop (PLL) on a predetermined sequence of
 - 10 observations from said block.
- 2.** The method for estimating the phase in a digital communication system according to claim 1 characterized by:
- 15
- executing a first phase locked loop (PLL) on said observations in order to generate a first intermediate value;
 - executing a second phase locked loop (PLL) on said observations in order to generate a second intermediate value;
 - 20 - combining said first and second intermediate values to generate a phase estimate.
- 3.** The method according to claim 2 characterized in that said first loop executes on a sequence of observations according to their chronological order of occurrence,
- 25 and that said second loop executes on the inverse sequence.
- 4.** The method according to claim 3 characterized in that said second phase locked loop (PLL) is initialized to the last value calculated by said first phase locked loop.
- 30
- 5.** The method according to claim 4 characterized in that it comprises the steps of:
- receiving and storing a block of observations Y_k , with k varying from 0 to n ;

- initializing a first phase locked loop from received observations Y_k ;
- executing said first phase locked loop according to the following formula:

$$\varphi_k = \varphi_{k-1} - \gamma F(Y_k, \varphi_{k-1}) \text{ with } k = 1 \text{ to } n$$

5 where F is a function adapted to the type of modulation considered

- initializing a second phase locked loop from observations Y_k , with k varying from n to 0 ;
- executing said second phase locked loop (PLL) according to the following formula:

10

$$\varphi'_k = \varphi'_{k+1} - \gamma F(Y_k, \varphi'_{k+1}) \text{ with } k = n-1 \text{ to } 0$$

- combining the results produced by said first and second loops to generate a phase estimate.

15

6. A method according to any of the preceding claims characterized in that the modulation is a binary phase shift keying (BPSK) modulation with a phase locked loop (PLL) defined by

20

$$\varphi_k = \varphi_{k-1} + \gamma \operatorname{Img}(y_k e^{-i\varphi(k-1)}) \operatorname{th}[L_k / 2 + 2 / \sigma^2 \operatorname{Re}(y_k e^{-i\varphi(k-1)})]$$

where:

25

th is the hyperbolic tangent operator,

Re is the operator referring to the real part of a complex number,

σ^2 is the noise variance;

and $L_k = \operatorname{Ln}[p(a_k = 1) / p(a_k = -1)]$,

30

and Ln is the natural logarithm, $p(a_k = 1)$ is the probability that symbol a_k is equal to $+1$ and $p(a_k = -1)$ is the probability that symbol a_k is equal to -1 .

7. The method according to claim 6 characterized in that said factor γ is realized by means of a second or higher order digital filter.

5 8. A phase locked loop device for a digital receiver comprising:

- means to receive and store blocks of observations;
- a first phase locked loop (PLL) for generating a first intermediate value;
- a second phase locked loop (PLL) for generating a second intermediate value;
- means to derive a phase estimate from said first and second intermediate values.

10

9. The phase locked loop device according to claim 8 characterized in that said first and second phase locked loops are realized according to the following formula:

15
$$\varphi_k = \varphi_{k-1} - \gamma F(Y_k, \varphi_{k-1}) \text{ with } k = 1 \text{ to } n$$

or

$$\varphi'_k = \varphi'_{k+1} - \gamma F(Y_k, \varphi'_{k+1}) \text{ with } k = n-1 \text{ to } 0$$

where F is a function adapted to the type of modulation considered.

20

10. The device according to claim 9 characterized in that the first value calculated by said second loop is determined by the last calculation made by said first phase locked loop.

25